

CS578 - SPEECH SIGNAL PROCESSING

LECTURE : QUASI-HARMONIC MODELS OF SPEECH

Yannis Pantazis
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(based on material from Prof. Stylianou)



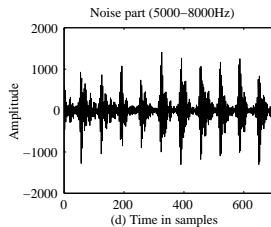
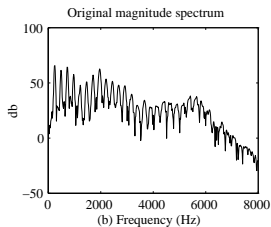
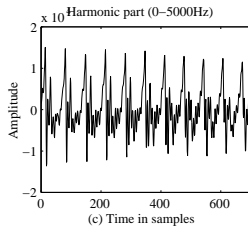
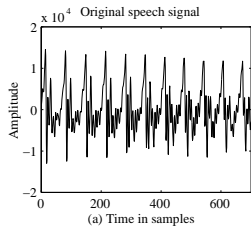
Speech Signal Processing Lab
Computer Science Department - University of Crete

2 November 2022

- 1 HARMONIC+NOISE MODELS
- 2 QUASI-HARMONIC MODEL - QHM
- 3 ITERATIVE QHM
- 4 ADAPTIVE QHM
- 5 EXTENSION OF AQHM
- 6 THANKS
- 7 REFERENCES

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MOTIVATION FOR HNM



BRIEF OVERVIEW OF HNM

- HNM (Stylianou 1995 [1]) is a pitch-synchronous harmonic plus noise representation of the speech signal.
- Speech spectrum is divided into a low and a high band delimited by the so-called *maximum voiced frequency*
- The *lower* band of the spectrum (below the maximum voiced frequency) is represented solely by harmonically related sine waves.
- The *upper* band is modeled as a noise component modulated by a time-domain amplitude envelope.
- HNM allows high-quality copy synthesis and prosodic modifications.

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- Harmonic part:

$$h(t) = \sum_{k=-K(t)}^{K(t)} A_k(t) e^{j2\pi k f_0(t) t}$$

where $A_k(t)$ and $f_0(t)$ are the instantaneous complex amplitude and real frequency, respectively

- Noise part:

$$n(t) = e(t) [v(\tau, t) \star g(t)]$$

where $e(t)$, $v(\tau, t)$, $g(t)$ are a time envelope, an estimation of the PSD (filter), and white gaussian noise, respectively

- Speech:

$$s(t) = h(t) + n(t)$$

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ESTIMATING SINUSOIDAL PARAMETERS

- Sinusoidal representation for a speech/signal frame:

$$x(t) = \left(\sum_{k=-K}^K a_k e^{j2\pi f_k t} \right) w(t)$$

- Parameter estimation methods:
 - FFT-based methods (i.e., QIFFT [Abe et al., 2004-05], [2], [3])
 - Least Squares (LS) method
- Frequency mismatch (eg, $\hat{f}_k := k\hat{f}_0$ in HNM):

$$\hat{f}_k = f_k + \eta_k$$

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QUASI-HARMONIC MODEL, QHM [5]

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- QHM (de Prony 1795, Laroche [4] (1989), Stylianou 1993, Pantazis [5] (2008, 2011)):

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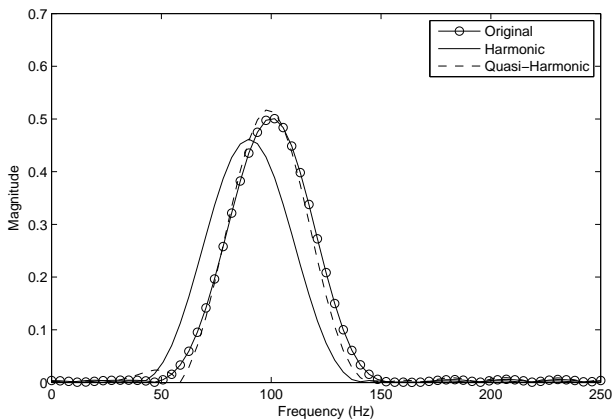
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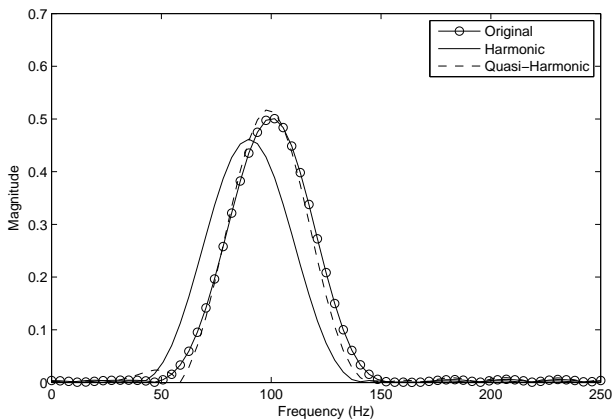
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- Inst. amplitude:

$$M_k(t) = |a_k + tb_k| = \sqrt{(a_k^R + tb_k^R)^2 + (a_k^I + tb_k^I)^2}$$

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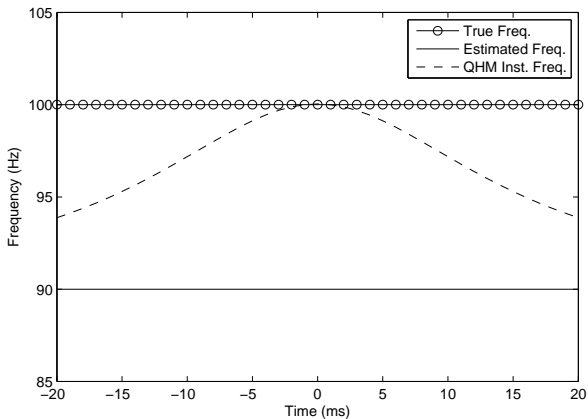
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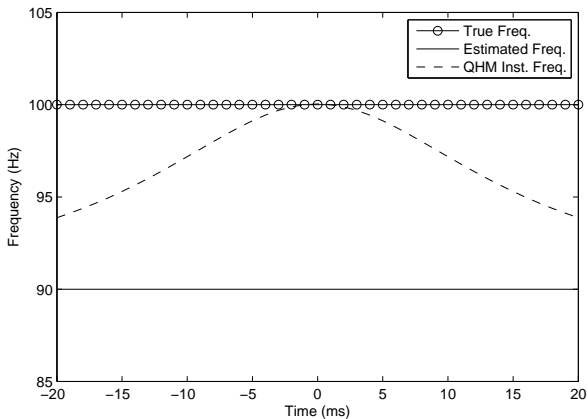
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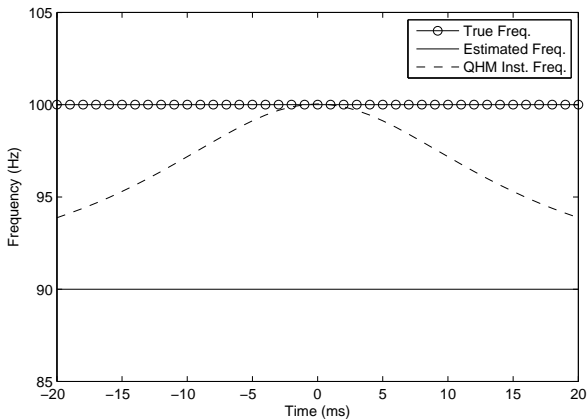
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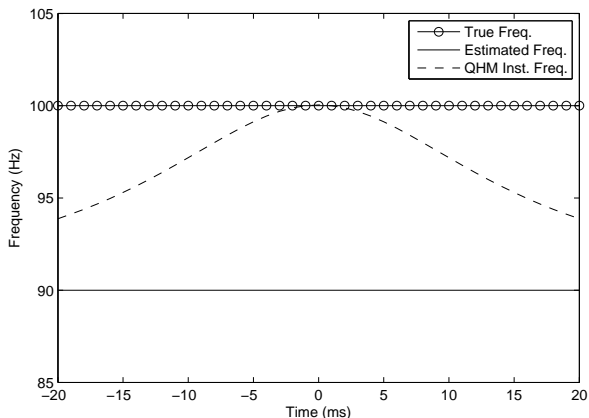
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A FEW DETAILS OF QHM

- QHM in the frequency domain:

$$X_k(f) = a_k W(f - \hat{f}_k) + j \frac{b_k}{2\pi} W'(f - \hat{f}_k)$$

- **Decomposition of b_k :** $b_k = \rho_{1,k} a_k + \rho_{2,k} j a_k$
- Then

$$X_k(f) = a_k \left[W(f - \hat{f}_k) - \frac{\rho_{2,k}}{2\pi} W'(f - \hat{f}_k) + j \frac{\rho_{1,k}}{2\pi} W'(f - \hat{f}_k) \right]$$

- and taking into account:

$$W\left(f - \hat{f}_k - \frac{\rho_{2,k}}{2\pi}\right) = W(f - \hat{f}_k) - \frac{\rho_{2,k}}{2\pi} W'(f - \hat{f}_k) + O(\rho_{2,k}^2 W''(f - \hat{f}_k))$$

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$$X_k(f) \approx a_k \left[W\left(f - \hat{f}_k - \frac{\rho_{2,k}}{2\pi}\right) + j \frac{\rho_{1,k}}{2\pi} W'(f - \hat{f}_k) \right]$$

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- Back to the time-domain:

$$x_k(t) \approx a_k \left[e^{j(2\pi\hat{f}_k + \rho_{2,k})t} + \rho_{1,k} t e^{j2\pi\hat{f}_k t} \right] w(t)$$

- Initially, we assumed:

$$x_k(t) = a_k \left[e^{j(2\pi(\hat{f}_k + \eta_k))t} \right] w(t)$$

- in other words, **it is suggested**:

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- However, this suggestion is conditional on the magnitude of $\rho_{2,k}$ and the value of term $W''(f)$ at f_k .
- Also, the correction term depends on the window main lobe's width

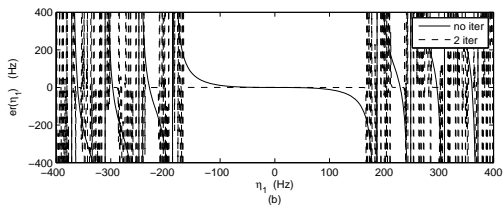
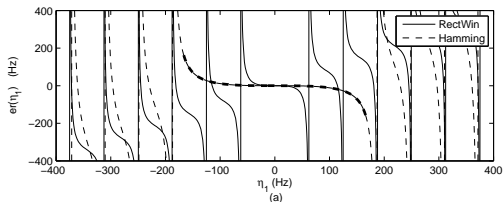
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SINGLE SINUSOID



- Iteratively, the bias can be removed when $|\eta| < B/3$, where B is the bandwidth of the squared analysis window.

QHM PARAMETER ESTIMATION

- Frame-by-frame analysis:

$$s[n], \quad n = -N, \dots, N$$

- QHM:

$$h[n] = \sum_{k=-K}^K (a_k + nb_k) e^{j2\pi f_k n / f_s}, \quad n = -N, \dots, N$$

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$$e[n] = w[n](s[n] - h[n]), \quad n = -N, \dots, N$$

QHM PARAMETER ESTIMATION

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QHM PARAMETER ESTIMATION

- Sum of squared error:

$$\begin{aligned}L(\mathbf{a}, \mathbf{b}) &= \sum_{n=-N}^N |e[n]|^2 \\ &= \sum_{n=-N}^N w[n]^2 |s[n] - (E_0(n, :) \mathbf{a} + E_1(n, :) \mathbf{b})|^2 \\ &= \left(\mathbf{s} - [E_0 | E_1] \begin{bmatrix} \mathbf{a} \\ \mathbf{b} \end{bmatrix} \right)^H W^2 \left(\mathbf{s} - [E_0 | E_1] \begin{bmatrix} \mathbf{a} \\ \mathbf{b} \end{bmatrix} \right)\end{aligned}$$

where

- $\mathbf{s} = [s[-N], \dots, s[N]]^T \in \mathbb{R}^{2N+1}$
- $\mathbf{a} = [a_{-K}, \dots, a_K]^T \in \mathbb{C}^{2K+1}$
- $\mathbf{b} = [b_{-K}, \dots, b_K]^T \in \mathbb{C}^{2K+1}$
- $(E_0)_{nk} = e^{j2\pi f_k n / f_s} \in \mathbb{C}^{(2N+1) \times (2K+1)}$
- $(E_1)_{nk} = n e^{j2\pi f_k n / f_s} \in \mathbb{C}^{(2N+1) \times (2K+1)}$

- Least Squares solution:

$$\begin{bmatrix} \hat{\mathbf{a}} \\ \hat{\mathbf{b}} \end{bmatrix} = \left([E_0|E_1]^H W^2 [E_0|E_1] \right)^{-1} [E_0|E_1]^H W^2 \mathbf{s}$$

- A necessary condition for well-posed estimation is

$$4K + 1 < 2N + 1$$

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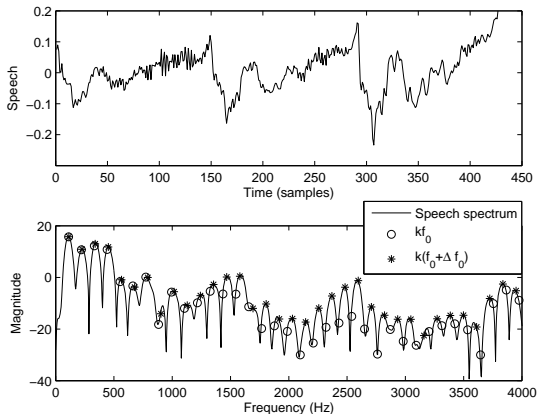
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- This frequency updating mechanism provides frequencies which can be used in the model iteratively and result in better parameter estimation (a_k, b_k)
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ITERATIVE QHM, IQHM [6]

HM versus iQHM in frequency estimation - speech signal:



ROBUSTNESS AGAINST ADDITIVE NOISE

- Signal contaminated by noise:

$$y(t) = \sum_{k=1}^4 a_k e^{j2\pi f_k t} + v(t)$$

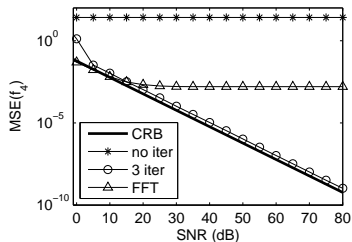
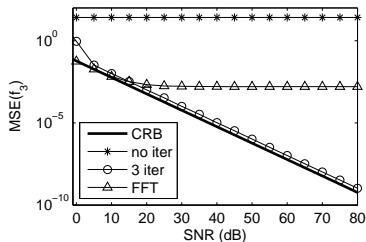
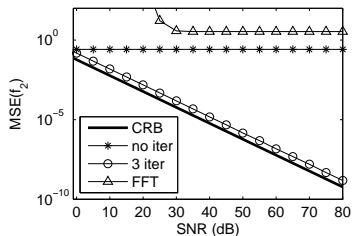
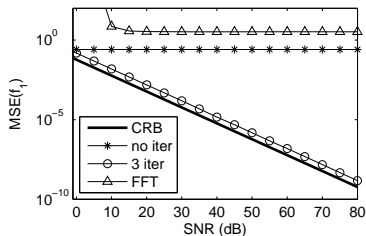
- Mean Squared Error (MSE):

$$MSE\{\hat{f}_k\} = \frac{1}{M} \sum_{i=1}^M |\hat{f}_k(i) - f_k|^2$$

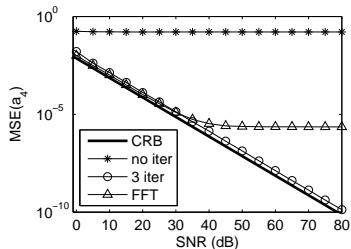
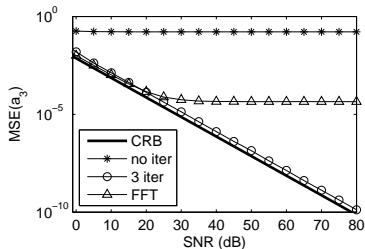
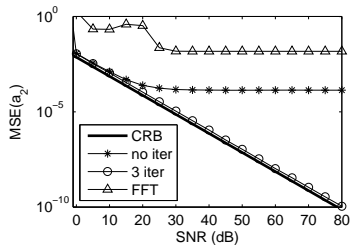
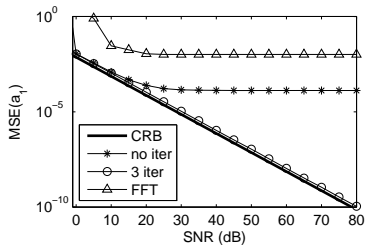
$$MSE\{\hat{a}_k\} = \frac{1}{M} \sum_{i=1}^M |\hat{a}_k(i) - a_k|^2$$

- Comparison with Cramer-Rao Bounds (CRB) and QIFFT (Abe et al. 2004)
- 10000 Monte Carlo simulations

MSE OF FREQUENCIES AS A FUNCTION OF SNR.



MSE OF AMPLITUDES AS A FUNCTION OF SNR.



- QHM has been shown to be closely related to:
 - Gauss-Newton frequency estimation method
 - Reassigned Spectrogram
 - AM-FM decomposition

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HIGH RESOLUTION AM-FM DECOMPOSITION

- AM-FM signal

$$y(t) = \sum_{k=1}^{K(t)} a_k(t) \cos(\phi_k(t)),$$

- Taylor series expansion of the instantaneous phase of k th component:

$$\phi_k(t) = 2\pi\zeta_k t + \sum_{i=0}^{\infty} \phi_{k,i} \frac{t^i}{i!}$$

- Instantaneous frequency of the k th component at $t = 0$:

$$\nu_k(0) = \zeta_k + \frac{\phi_{k,1}}{2\pi}$$

- ... and previously we had:

$$F_k(0) = f_k + \frac{\rho_{2,k}}{2\pi}$$

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FROM QHM TO ADAPTIVE QHM, AQHM [7]

- QHM (**stationarity assumption**):

$$x(t) = \left(\sum_{k=-K}^K (a_k + tb_k) e^{2\pi j f_k t} \right) w(t)$$

- Adaptive QHM (aQHM):

$$x(t) = \left(\sum_{k=-K}^K (a_k + tb_k) e^{j\tilde{\phi}_k(t)} \right) w(t)$$

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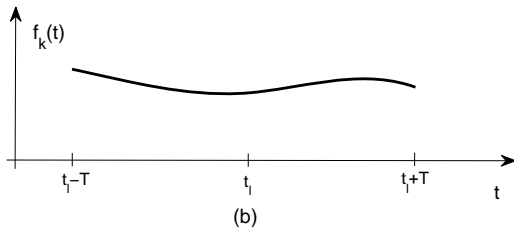
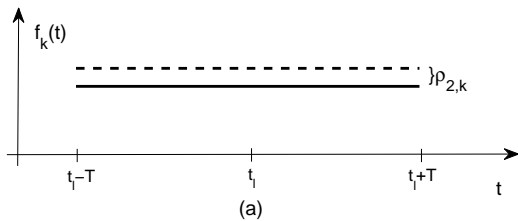
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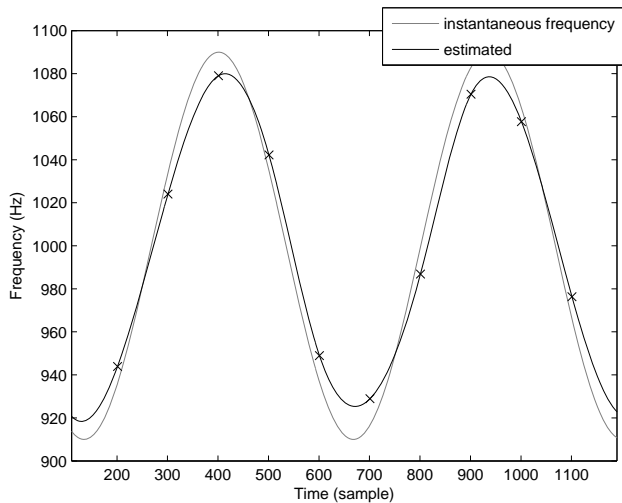
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FROM QHM TO AQHM; GRAPHICALLY

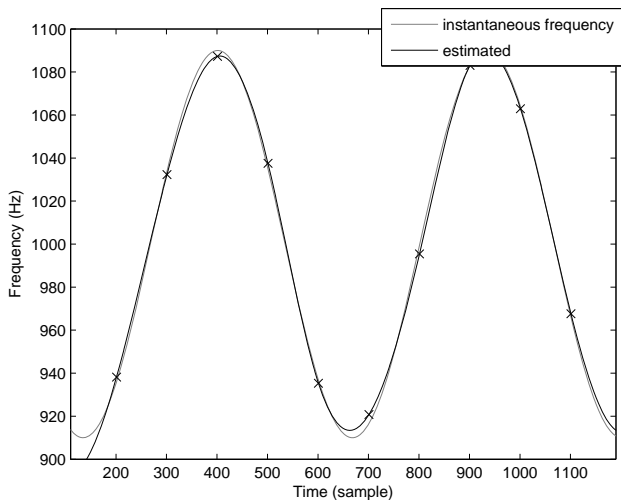


- One sample: no interpolation between estimations
- Higher rates (i.e., 5ms, 10ms): Interpolation between estimates is required:
 - Amplitudes are linearly interpolated
 - Frequencies are interpolated with splines
 - Phases are interpolated by integration of instantaneous frequency

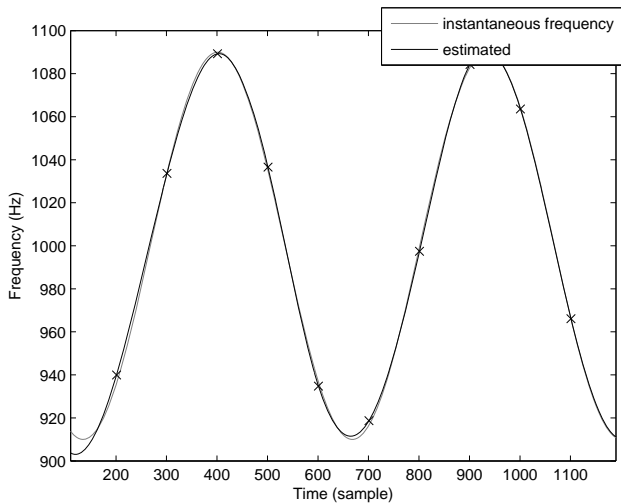
EXAMPLE OF ESTIMATION IN AQHM: NO ITERATION (QHM)



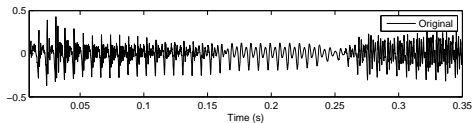
EXAMPLE OF ESTIMATION IN AQHM: ONE ITERATION



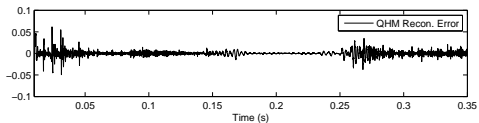
EXAMPLE OF ESTIMATION IN AQHM: TWO ITERATIONS



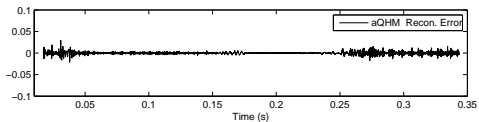
RECONSTRUCTION ERRORS WITH QHM, AQHM, SM



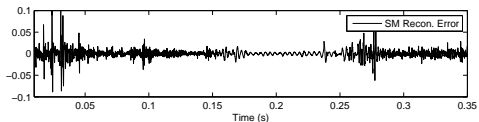
(a)



(b)



(c)



(d)

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EXTENDED AQHM

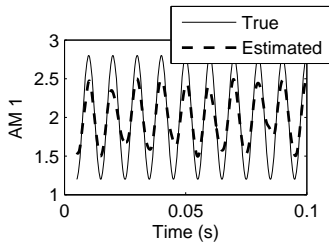
- Recall aQHM:

$$x(t) = \sum_{k=-K}^K (a_k + tb_k) e^{j\tilde{\phi}_k(t)}$$

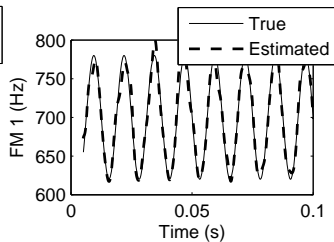
- Extended aQHM:

$$x(t) = \sum_{k=-K}^K (a_k + tb_k) \tilde{\alpha}(t) e^{j\tilde{\phi}_k(t)}$$

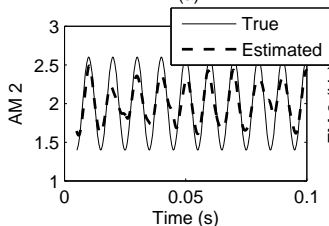
AM-FM MODELING: AQHM



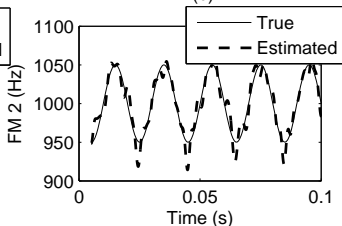
(a)



(b)

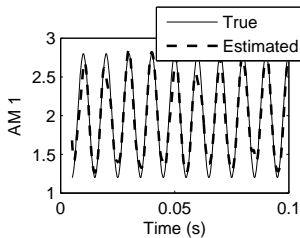


(c)

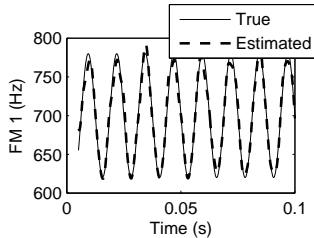


(d)

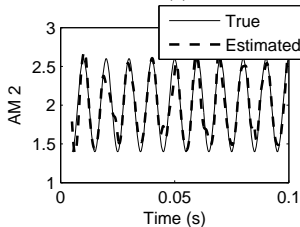
AM-FM MODELING: EXTENDED AQHM



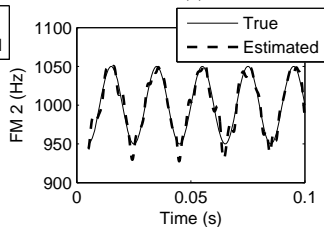
(a)



(b)

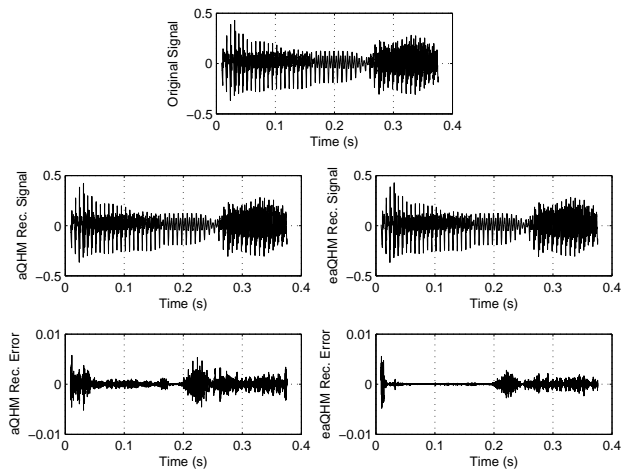


(c)



(d)

COMPARING ADAPTIVE MODELS



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THANK YOU
for your attention

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