CS578- Speech Signal Processing Lecture 5: Sinusoidal modeling and modifications

Yannis Stylianou



University of Crete, Computer Science Dept., Multimedia Informatics Lab yannis@csd.uoc.gr

Univ. of Crete

▲ロ ▶ ▲周 ▶ ▲ 国 ▶ ▲ 国 ▶ ● の Q @

OUTLINE

- **1** Sinusoidal Speech Model Voiced Speech Unvoiced Speech The Analysis System **3** Synthesis Linear Amplitude Interpolation Cubic Phase Interpolation 6 Shape Invariant Time-Scale Modifications The Model Parameters Estimation Synthesis Sound Examples
 - ▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへ⊙

Source-Filter[1]

• Source:

$$u(t) = Re \sum_{k=1}^{K(t)} \alpha_k(t) \exp \left[j \phi_k(t) \right]$$

where:

$$\phi_k(t) = \int_0^t \Omega_k(\sigma) d\sigma + \phi_k$$

• Filter: $h(t, \tau)$ with Fourier Transform (FT):

 $H(t, \Omega) = M(t, \Omega) \exp [j\Phi(t, \Omega)]$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

Source-Filter[1]

• Source:

$$u(t) = Re \sum_{k=1}^{K(t)} \alpha_k(t) \exp \left[j \phi_k(t) \right]$$

where:

$$\phi_k(t) = \int_0^t \Omega_k(\sigma) d\sigma + \phi_k$$

• Filter: $h(t, \tau)$ with Fourier Transform (FT):

 $H(t, \Omega) = M(t, \Omega) \exp [j\Phi(t, \Omega)]$

◆□▶ ◆□▶ ◆三▶ ◆三▶ 三三 のへぐ

OUTPUT SPEECH

$$s(t) = Re \sum_{k=1}^{K(t)} A_k(t) \exp \left[j\theta_k(t)\right]$$

where:

$$\begin{array}{lll} A_k(t) &=& \alpha_k(t) M\left[t, \Omega_k(t)\right] \\ \theta_k(t) &=& \phi_k(t) + \Phi\left[t, \Omega_k(t)\right] \\ &=& \int_0^t \Omega_k(\sigma) d\sigma + \Phi\left[t, \Omega_k(t)\right] + \phi_k \end{array}$$

◆□▶ ◆□▶ ◆ 臣▶ ◆ 臣▶ ○ 臣 ○ の Q @

OUTLINE

- SINUSOIDAL SPEECH MODEL
- **2** Estimation of Sinewave Parameters
 - Voiced Speech
 - Unvoiced Speech
 - The Analysis System
- **3** Synthesis
 - Linear Amplitude Interpolation
 - Cubic Phase Interpolation
- 4 Examples
- **5** Sound Examples
- 6 Shape Invariant Time-Scale Modifications

- The Model
- Parameters Estimation
- Synthesis
- Sound Examples
- 7 Shape Invariant Pitch Modifications
- 8 Acknowledgments
- **9** References

FRAME-BY-FRAME ANALYSIS



◆□▶ ◆□▶ ◆三▶ ◆三▶ ○三 の々で

STATIONARITY ASSUMPTION

We assume stationarity inside the analysis window:

$$egin{array}{rcl} A_k^l(t) &=& A_k^l\ \Omega_k^l(t) &=& \Omega_k^l \end{array}$$

which leads to:

$$heta_k^{\prime}(t) = \Omega_k^{\prime}(t-t_l) + heta_k^{\prime}$$

and to:

$$s(t) = \sum_{k=1}^{K'} A_k^{\prime} \exp\left(j\theta_k^{\prime}\right) \exp\left[j\Omega_k^{\prime}(t-t_l)\right] \quad t_l - \frac{T}{2} \le t \le t_l + \frac{T}{2}$$

◆□▶ ◆□▶ ◆∃▶ ◆∃▶ = ● ● ●

We assume stationarity inside the analysis window:

$$egin{array}{rcl} A_k^l(t) &=& A_k^l\ \Omega_k^l(t) &=& \Omega_k^l \end{array}$$

which leads to:

$$heta_k^{\prime}(t) = \Omega_k^{\prime}(t-t_l) + heta_k^{\prime}$$

and to:

$$s(t) = \sum_{k=1}^{K'} A_k^l \exp\left(j\theta_k^l\right) \exp\left[j\Omega_k^l(t-t_l)\right] \quad t_l - \frac{T}{2} \le t \le t_l + \frac{T}{2}$$

We assume stationarity inside the analysis window:

$$\begin{array}{rcl} A_k^l(t) &=& A_k^l\\ \Omega_k^l(t) &=& \Omega_k^l \end{array}$$

which leads to:

$$heta_k^l(t) = \Omega_k^l(t-t_l) + heta_k^l$$

and to:

$$s(t) = \sum_{k=1}^{K'} A_k^l \exp\left(j\theta_k^l
ight) \exp\left[j\Omega_k^l(t-t_l)
ight] \quad t_l - \frac{T}{2} \le t \le t_l + \frac{T}{2}$$

(ロ) (型) (E) (E) (E) (O)(C)

Steps to discrete time formula:

- Time shift: $t' = t t_l$
- Convert to discrete time:

$$s[n] = \sum_{k=1}^{K'} A'_k \exp\left(j\theta'_k\right) \exp\left(j\omega'_k n\right) \quad -\frac{N_w - 1}{2} \le n \le \frac{N_w - 1}{2}$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

MEAN-SQUARED ERROR

Given the original measured waveform, y[n] and the synthetic speech waveform, s[n], estimate the unknown parameters A_k^l , ω_k^l , and θ_k^l by minimizing the MSE criterion:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n] - s[n]|^2$$

which can be written as:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n]|^2 + N_w \sum_{k=1}^{K'} \left(\left| Y(\omega_k') - \gamma_k' \right|^2 - |Y(\omega_k')|^2 \right)$$

which can be reduced further to:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n]|^2 - N_w \sum_{k=1}^{K'} |Y(\omega'_k)|^2$$

MEAN-SQUARED ERROR

Given the original measured waveform, y[n] and the synthetic speech waveform, s[n], estimate the unknown parameters A_k^l , ω_k^l , and θ_k^l by minimizing the MSE criterion:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n] - s[n]|^2$$

which can be written as:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n]|^2 + N_w \sum_{k=1}^{K'} \left(\left| Y(\omega_k') - \gamma_k' \right|^2 - |Y(\omega_k')|^2 \right)$$

which can be reduced further to:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n]|^2 - N_w \sum_{k=1}^{K'} |Y(\omega'_k)|^2$$

MEAN-SQUARED ERROR

Given the original measured waveform, y[n] and the synthetic speech waveform, s[n], estimate the unknown parameters A_k^l , ω_k^l , and θ_k^l by minimizing the MSE criterion:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n] - s[n]|^2$$

which can be written as:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n]|^2 + N_w \sum_{k=1}^{K'} \left(\left| Y(\omega_k') - \gamma_k' \right|^2 - |Y(\omega_k')|^2 \right)$$

which can be reduced further to:

$$\epsilon' = \sum_{n=-(N_w-1)/2}^{n=(N_w-1)/2} |y[n]|^2 - N_w \sum_{k=1}^{K'} |Y(\omega'_k)|^2$$

- Karhunen-Loève expansion allows constructing a random process from harmonic sinusoids with uncorrelated complex amplitudes.
- Estimated power spectrum should not vary "too much" over consecutive frequencies.

Following the above necessary constraints, for unvoiced speech, and for a window width to be *at least* 20ms, an 100 Hz harmonic structure provides good results.

▲ロ ▶ ▲周 ▶ ▲ 国 ▶ ▲ 国 ▶ ● の Q @

- Karhunen-Loève expansion allows constructing a random process from harmonic sinusoids with uncorrelated complex amplitudes.
- Estimated power spectrum should not vary "too much" over consecutive frequencies.

Following the above necessary constraints, for unvoiced speech, and for a window width to be *at least* 20ms, an 100 Hz harmonic structure provides good results.

▲□▶ ▲□▶ ▲□▶ ▲□▶ ▲□ ● ● ●

- Karhunen-Loève expansion allows constructing a random process from harmonic sinusoids with uncorrelated complex amplitudes.
- Estimated power spectrum should not vary "too much" over consecutive frequencies.

Following the above necessary constraints, for unvoiced speech, and for a window width to be *at least* 20ms, an 100 Hz harmonic structure provides good results.

▲□▶ ▲□▶ ▲□▶ ▲□▶ ▲□ ● ● ●

- Karhunen-Loève expansion allows constructing a random process from harmonic sinusoids with uncorrelated complex amplitudes.
- Estimated power spectrum should not vary "too much" over consecutive frequencies.

Following the above necessary constraints, for unvoiced speech, and for a window width to be *at least* 20ms, an 100 Hz harmonic structure provides good results.

▲□▶ ▲□▶ ▲□▶ ▲□▶ ▲□ ● ● ●



◆□▶ ◆□▶ ◆三▶ ◆三▶ ○三 の々で

- Window width be 2.5 times the average pitch period or 20 ms, whichever is larger.
- Use Hamming window, normalized to one:

$$\sum_{n=-\infty}^{\infty} w[n] = 1$$

- Use zero padding to get enough samples of the underlying spectrum (i.e., 1024-point FFT)
- Remove linear phase offset
- Refine your frequency estimates

- Window width be 2.5 times the average pitch period or 20 ms, whichever is larger.
- Use Hamming window, normalized to one:

$$\sum_{n=-\infty}^{\infty} w[n] = 1$$

- Use zero padding to get enough samples of the underlying spectrum (i.e., 1024-point FFT)
- Remove linear phase offset
- Refine your frequency estimates

- Window width be 2.5 times the average pitch period or 20 ms, whichever is larger.
- Use Hamming window, normalized to one:

$$\sum_{n=-\infty}^{\infty} w[n] = 1$$

- Use zero padding to get enough samples of the underlying spectrum (i.e., 1024-point FFT)
- Remove linear phase offset
- Refine your frequency estimates

- Window width be 2.5 times the average pitch period or 20 ms, whichever is larger.
- Use Hamming window, normalized to one:

$$\sum_{n=-\infty}^{\infty} w[n] = 1$$

- Use zero padding to get enough samples of the underlying spectrum (i.e., 1024-point FFT)
- Remove linear phase offset
- Refine your frequency estimates

- Window width be 2.5 times the average pitch period or 20 ms, whichever is larger.
- Use Hamming window, normalized to one:

$$\sum_{n=-\infty}^{\infty} w[n] = 1$$

- Use zero padding to get enough samples of the underlying spectrum (i.e., 1024-point FFT)
- Remove linear phase offset
- Refine your frequency estimates

SHOWING THE PROCESS ...



▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

BLOCK DIAGRAM OF THE ANALYSIS SYSTEM



▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

OUTLINE

- **1** Sinusoidal Speech Model
- **2** Estimation of Sinewave Parameters
 - Voiced Speech
 - Unvoiced Speech
 - The Analysis System
- **3** Synthesis
 - Linear Amplitude Interpolation
 - Cubic Phase Interpolation
- 4 Examples
- **5** Sound Examples
- 6 Shape Invariant Time-Scale Modifications

▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

- The Model
- Parameters Estimation
- Synthesis
- Sound Examples
- 7 Shape Invariant Pitch Modifications
- 8 ACKNOWLEDGMENTS
- **9** References

PROBLEM OF FREQUENCY MATCHING



▲□▶ ▲□▶ ▲ □▶ ▲ □▶ □ のへぐ

FRAME-TO-FRAME PEAK MATCHING



・ロト・日本・日本・日本・日本・日本

THE BIRTH/DEATH PROCESS





A BIRTH/DEATH PROCESS IN SPEECH



Why not to estimate the original speech waveform on the *I*th frame, directly as:

$$s[n] = \sum_{k=1}^{K'} A'_k \cos(n\omega'_k + \theta'_k), \quad n = 0, 1, 2, \cdots, L-1$$

◆□▶ ◆□▶ ◆ 臣▶ ◆ 臣▶ ○ 臣 ○ の Q @

A SIMPLE SOLUTION: OLA



▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

Linear Interpolation:

$$A'_{k}[n] = A'_{k} + \left(A'^{+1}_{k} - A'_{k}\right)\left(\frac{n}{L}\right) \quad n = 0, 1, 2, \cdots, L-1$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

PHASE WRAPPED



CUBIC PHASE MODEL

$$\theta(t) = \zeta + \gamma t + \alpha t^2 + \beta t^3$$

(ロ)、(型)、(E)、(E)、 E) の(()
Assuming that vocal tract is slowly varying, and since:

$$\theta(t) = \int_0^t \Omega(\sigma) d\sigma + \phi + \Phi[t, \Omega(t)]$$

 $\dot{ heta}(t)pprox \Omega(t)$

So:

$$\dot{ heta}' \approx \Omega' \ \dot{ heta}'^{+1} \approx \Omega'^{+1}$$

◆□▶ ◆□▶ ◆ □▶ ◆ □▶ ● ● ● ●

Assuming that vocal tract is slowly varying, and since:

$$\theta(t) = \int_0^t \Omega(\sigma) d\sigma + \phi + \Phi[t, \Omega(t)]$$

 $\dot{ heta}(t) pprox \Omega(t)$

So:

 $\dot{ heta}' \approx \Omega' \ \dot{ heta}'^{+1} \approx \Omega'^{+1}$

Assuming that vocal tract is slowly varying, and since:

$$egin{aligned} & heta(t) = \int_0^t \Omega(\sigma) d\sigma + \phi + \Phi[t,\Omega(t)] \ &\dot{ heta}(t) pprox \Omega(t) \end{aligned}$$

So:

$$egin{array}{ccc} \dot{ heta}^{\prime} &pprox & \Omega^{\prime} \ \dot{ heta}^{\prime+1} &pprox & \Omega^{\prime+1} \end{array}$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

$$\begin{aligned} \theta(0) &= \theta' \\ \dot{\theta}(0) &= \Omega' \\ \theta(T) &= \theta'^{l+1} + 2\pi M \\ \dot{\theta}(T) &= \Omega'^{l+1} \end{aligned}$$

▲□▶ ▲□▶ ▲ □▶ ▲ □▶ □ のへぐ

and ... five unknowns (don't forget M) We need one more constraint!

$$\begin{aligned} \theta(0) &= \theta' \\ \dot{\theta}(0) &= \Omega' \\ \theta(T) &= \theta'^{l+1} + 2\pi M \\ \dot{\theta}(T) &= \Omega'^{l+1} \end{aligned}$$

◆□▶ ◆□▶ ◆三▶ ◆三▶ 三三 のへぐ

and ... five unknowns (don't forget M) We need one more constraint!

$$\begin{aligned} \theta(0) &= \theta' \\ \dot{\theta}(0) &= \Omega' \\ \theta(T) &= \theta'^{l+1} + 2\pi M \\ \dot{\theta}(T) &= \Omega'^{l+1} \end{aligned}$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三 のへぐ

and ... five unknowns (don't forget M) We need one more constraint!

$$\begin{aligned} \theta(0) &= \theta' \\ \dot{\theta}(0) &= \Omega' \\ \theta(T) &= \theta'^{l+1} + 2\pi M \\ \dot{\theta}(T) &= \Omega'^{l+1} \end{aligned}$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三 のへぐ

and ... five unknowns (don't forget M) We need one more constraint!

How to choose M



• Find M that minimizes the criterion:

$$f(M) = \int_0^T \left[\ddot{\theta}(t;M)\right]^2 dt$$

• Using continuous variable:

$$x^* = \frac{1}{2\pi} \left[\left(\theta^l + \Omega^l T - \theta^{l+1} \right) + \left(\Omega^{l+1} - \Omega^l \right) \frac{T}{2} \right]$$

◆□▶ ◆□▶ ◆ 臣▶ ◆ 臣▶ ○ 臣 ○ の Q @

• *M*^{*} is the nearest integer to *x*^{*}

• Find M that minimizes the criterion:

$$f(M) = \int_0^T \left[\ddot{\theta}(t;M) \right]^2 dt$$

• Using continuous variable:

$$x^* = \frac{1}{2\pi} \left[(\theta' + \Omega'T - \theta'^{+1}) + (\Omega'^{+1} - \Omega')\frac{T}{2} \right]$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

• *M*^{*} is the nearest integer to *x*^{*}

• Find M that minimizes the criterion:

$$f(M) = \int_0^T \left[\ddot{\theta}(t;M) \right]^2 dt$$

• Using continuous variable:

$$x^* = rac{1}{2\pi} \left[(heta^l + \Omega^l T - heta^{l+1}) + (\Omega^{l+1} - \Omega^l) rac{T}{2}
ight]$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

• M^* is the nearest integer to x^*

BLOCK DIAGRAM OF THE SYNTHESIS SYSTEM



▲□▶ ▲圖▶ ▲匡▶ ▲匡▶ ― 匡 … のへで

OUTLINE

- **1** Sinusoidal Speech Model
- **2** Estimation of Sinewave Parameters
 - Voiced Speech
 - Unvoiced Speech
 - The Analysis System
- **3** Synthesis
 - Linear Amplitude Interpolation
 - Cubic Phase Interpolation

4 Examples

- **5** Sound Examples
- 6 Shape Invariant Time-Scale Modifications

▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

- The Model
- Parameters Estimation
- Synthesis
- Sound Examples
- 7 Shape Invariant Pitch Modifications
- 8 Acknowledgments
- **9** References

RECONSTRUCTION EXAMPLE



◆□▶ ◆□▶ ◆三▶ ◆三▶ 三三 - のへで

RECONSTRUCTION EXAMPLE



◆□▶ ◆□▶ ◆三▶ ◆三▶ 三三 - のへで

MAGNITUDE-ONLY RECONSTRUCTION EXAMPLE



▲□▶ ▲□▶ ▲臣▶ ★臣▶ = 臣 = のへで

OUTLINE

- 1 Sinusoidal Speech Model
- **2** Estimation of Sinewave Parameters
 - Voiced Speech
 - Unvoiced Speech
 - The Analysis System
- **3** Synthesis
 - Linear Amplitude Interpolation
 - Cubic Phase Interpolation
 - Examples
- **5** Sound Examples
 - 6 Shape Invariant Time-Scale Modifications

▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

- The Model
- Parameters Estimation
- Synthesis
- Sound Examples
- 7 Shape Invariant Pitch Modifications
- 8 Acknowledgments
- **9** References

Sound Examples





OUTLINE

Voiced Speech Unvoiced Speech The Analysis System **3** Synthesis Linear Amplitude Interpolation Cubic Phase Interpolation **5** Sound Examples **6** Shape Invariant Time-Scale Modifications The Model Parameters Estimation Synthesis Sound Examples

▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

- 8 Acknowledgments
- 9 REFERENCES

EXCITATION MODEL

We have seen that:

$$u(t) = \sum_{k=1}^{K(t)} \alpha_k(t) \exp\left[j\phi_k(t)\right]$$

where:

$$\phi_k(t) = \int_0^t \Omega_k(\sigma) d\sigma + \phi_k$$

Assuming voiced speech and constant frequency in the analysis window, then:

$$u(t) = \sum_{k=1}^{K(t)} \alpha_k(t) \exp\left[j(t-t_0)\Omega_k\right] \quad t \in [0, T]$$

◆□ ▶ ◆□ ▶ ◆ 臣 ▶ ◆ 臣 ▶ ○ 臣 ○ の Q @

We have seen that:

$$u(t) = \sum_{k=1}^{K(t)} \alpha_k(t) \exp\left[j\phi_k(t)\right]$$

where:

$$\phi_k(t) = \int_0^t \Omega_k(\sigma) d\sigma + \phi_k$$

Assuming voiced speech and constant frequency in the analysis window, then:

$$u(t) = \sum_{k=1}^{K(t)} \alpha_k(t) \exp\left[j(t-t_0)\Omega_k\right] \quad t \in [0, T]$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

Speech model[2]

Then:

$$s[n] = \sum_{k=1}^{K(t)} A_k(t) \cos \left[\theta_k(t)\right]$$

where:

$$egin{array}{rcl} A_k(t)&=&lpha_k(t)M_k(t)\ heta_k(t)&=&\phi_k(t)+\Phi_k(t) \end{array}$$

Therefore:

$$\Phi_k(t) = \theta_k(t) - (t - t_0)\Omega_k$$

Let's t represent the original articulation rate and t' the transformed rate:

$t' = \rho t$

Given the source/filter model:

- System parameters are time-scaled
- Excitation parameters (phase) are scaled in such a way to maintain fundamental frequency.

Let's t represent the original articulation rate and t' the transformed rate:

$$t' = \rho t$$

Given the source/filter model:

- System parameters are time-scaled
- Excitation parameters (phase) are scaled in such a way to maintain fundamental frequency.

▲□▶ ▲□▶ ▲□▶ ▲□▶ ■ ●の00

ONSET-TIME MODEL FOR TIME-SCALE



◆□▶ ◆□▶ ◆ 臣▶ ◆ 臣▶ 三臣 - ∽ � � �

EXCITATION FUNCTION IN t'

• Time-scaled pitch period:

$$\tilde{P}(t') = P(t'\rho^{-1})$$

Modified excitation function

$$\tilde{u}(t') = \sum_{k=1}^{K(t)} \tilde{\alpha}_k(t') \exp\left[j\tilde{\phi}_k(t')\right]$$

where:

$$\tilde{\phi}_{k}(t') = (t' \rho^{-1} - t_{0}') \Omega_{k}$$

and

$$\tilde{\alpha}_k(t') = \alpha_k(t'\rho^{-1})$$

▲□▶ ▲□▶ ▲ 三▶ ▲ 三▶ 三三 - のへぐ

System function parameters in t'

$$egin{array}{rcl} ilde{M}_k(t') &=& M_k(t'
ho^{-1}) \ ilde{\Phi}_k(t') &=& \Phi_k(t'
ho^{-1}) \end{array}$$

◆□▶ ◆□▶ ◆臣▶ ◆臣▶ 臣 の�?

Waveform in t'

$$ilde{s}(t') = \sum_{k=1}^{\mathcal{K}(t)} ilde{A}_k(t') \exp\left[j ilde{ heta}_k(t')
ight]$$

where

$$egin{array}{rcl} ilde{\mathcal{A}}_k(t') &=& ilde{lpha}_k(t') ilde{\mathcal{H}}_k(t') \ ilde{ heta}_k(t') &=& ilde{\phi}_k(t') + ilde{\Phi}_k(t') \end{array}$$

(ロ)、(型)、(E)、(E)、 E) の(()

ONSET TIMES ESTIMATION



◆□▶ ◆□▶ ◆ □▶ ◆ □▶ ○ □ ○ ○ ○ ○

Let's assume that the onset time $n_o(l)$ for the l^{th} frame is known, then:

$$\phi_k^I = \hat{n}_o(I)\omega_k^I$$

where $\hat{n}_o(l) = n_o(l) - lL$. Then, the system phase is estimated as:

$$\tilde{\Phi}_k^I = \theta_k^I - \phi_k^I$$

◆□▶ ◆□▶ ◆三▶ ◆三▶ 三三 のへぐ

Let's assume that the onset time $n_o(I)$ for the I^{th} frame is known, then:

$$\phi_k^l = \hat{n}_o(l)\omega_k^l$$

where $\hat{n}_o(l) = n_o(l) - lL$. Then, the system phase is estimated as:

$$\tilde{\Phi}_k^I = \theta_k^I - \phi_k^I$$

◆□▶ ◆□▶ ◆三▶ ◆三▶ 三三 のへぐ

Let's assume we know the onset time in the previous frame l-1, then the current onset time in t', is given by:

$$n'_{o}(l) = n'_{o}(l-1) + J'P'$$

and then:

$$\tilde{\phi}_{k}^{I} = (n_{o}^{'}(I) - IL^{'})\omega_{k}^{I}$$

◆□▶ ◆□▶ ◆三▶ ◆三▶ 三三 のへぐ

where $L' = \rho L$

Synthesis is performed in the same way as if no modification is applied:

▲□▶ ▲□▶ ▲ □▶ ▲ □▶ □ のへぐ

- Linear interpolation for amplitudes
- Cubic interpolation for phases

BLOCK DIAGRAM FOR ANALYSIS/SYNTHESIS FOR TIME-SCALE MODIFICATION



EXAMPLE OF TIME-SCALE MODIFICATION



▲□▶ ▲□▶ ▲臣▶ ★臣▶ = 臣 = のへで

Sound Examples


OUTLINE

- **1** Sinusoidal Speech Model
- **2** Estimation of Sinewave Parameters
 - Voiced Speech
 - Unvoiced Speech
 - The Analysis System
- **3** Synthesis
 - Linear Amplitude Interpolation
 - Cubic Phase Interpolation
- 4 Examples
- **5** Sound Examples
- 6 Shape Invariant Time-Scale Modifications

▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

- The Model
- Parameters Estimation
- Synthesis
- Sound Examples
- **7** Shape Invariant Pitch Modifications
- **8** Acknowledgments
- **9** References

Paper:

T. F. Quatieri and R. J. McAulay: Shape Invariant Time-Scale and Pitch Modification of Speech IEEE Trans. Acoust., Speech, Signal Processing, Vol.40, No.3, pp 497-510, March 1992

▲□▶ ▲□▶ ▲□▶ ▲□▶ ▲□ ● ● ●

OUTLINE

- **1** Sinusoidal Speech Model
- **2** Estimation of Sinewave Parameters
 - Voiced Speech
 - Unvoiced Speech
 - The Analysis System
- **3** Synthesis
 - Linear Amplitude Interpolation
 - Cubic Phase Interpolation
- 4 Examples
- **5** Sound Examples
- 6 Shape Invariant Time-Scale Modifications

▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

- The Model
- Parameters Estimation
- Synthesis
- Sound Examples
- 7 Shape Invariant Pitch Modifications
- 8 Acknowledgments
- 9 REFERENCES

Most, if not all, figures in this lecture are coming from the book:

T. F. Quatieri: Discrete-Time Speech Signal Processing, principles and practice 2002, Prentice Hall

▲□▶ ▲□▶ ▲□▶ ▲□▶ ▲□ ● ● ●

and have been used after permission from Prentice Hall

OUTLINE

- **1** Sinusoidal Speech Model
- **2** Estimation of Sinewave Parameters
 - Voiced Speech
 - Unvoiced Speech
 - The Analysis System
- **3** Synthesis
 - Linear Amplitude Interpolation
 - Cubic Phase Interpolation
- 4 Examples
- **5** Sound Examples
- 6 Shape Invariant Time-Scale Modifications

▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 のへで

- The Model
- Parameters Estimation
- Synthesis
- Sound Examples
- 7 Shape Invariant Pitch Modifications
- **8** Acknowledgments
- 9 References



R. J. McAulay and T. F. Quatieri, "Speech analysis/synthesis based on a sinusoidal representation," *IEEE Trans. Acoust., Speech, Signal Processing*, vol. ASSP-34, pp. 744–754, Aug 1986.



T. F. Quatieri and R. J. McAulay, "Shape Invariant Time-Scale and Pitch Modification of Speech," *IEEE Trans. Acoust., Speech, Signal Processing*, vol. ASSP-40, pp. 497–510, March 1992.

◆□▶ ◆□▶ ◆三▶ ◆三▶ 三三 のへぐ

ふして 山田 ふぼやえばや 山下